

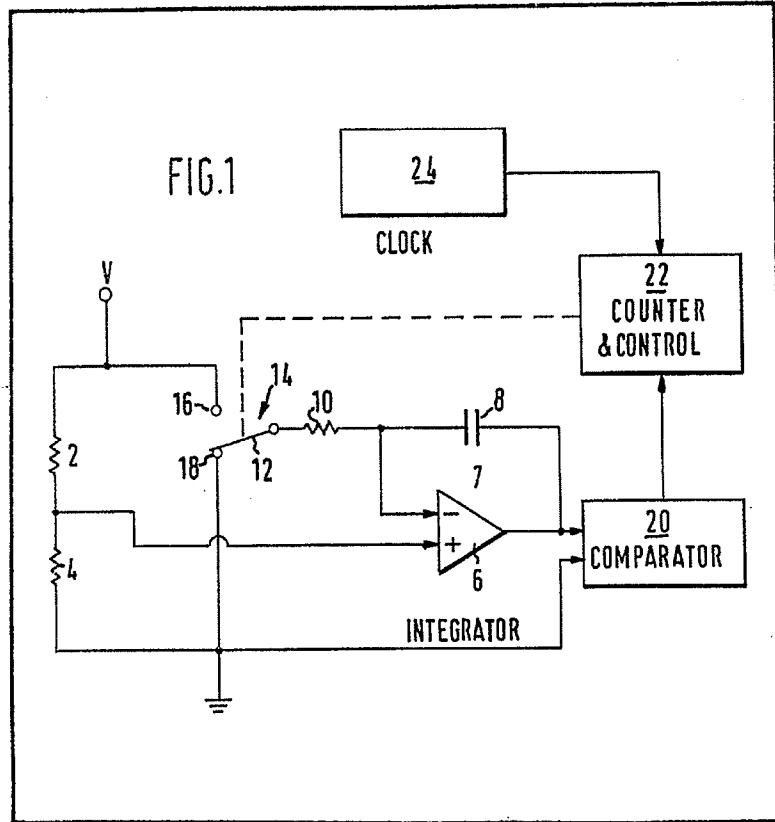
(12) UK Patent Application (19) GB (11) 2 076 547 A

(21) Application No 8112542
(22) Date of filing 23 Apr 1981
(30) Priority data
(31) 150412
(32) 16 May 1980
(33) United States of America (US)
(43) Application published 2 Dec 1981
(51) INT CL³
G01R 27/14
(52) Domestic classification G1U JB6
(56) Documents cited GB 1500695
(58) Field of search G1U
(71) Applicant Honeywell Inc.,
Honeywell Plaza,
Minneapolis, Minnesota
55408, United States of
America
(72) Inventor Frank Lynch
(74) Agent John Riddle & M. G.
Harman,
Honeywell Control
Systems Ltd., Patent
Department, Charles
Square, Bracknell, Berks.

(54) Resistance Measuring Circuit

(57) A resistance measuring circuit comprises a reference resistance 2 and the resistance to be measured 4, connected in series between a constant voltage source V and ground, an integrator being connected as

shown in Figure 1 to form a dual range voltmeter. Integration is performed from a threshold level V_T for a fixed time interval T_1 , the time T_2 for the integrator to return to the threshold level V_T is then measured. The ratio of the times T_1/T_2 is then proportional to the ratio of the reference and unknown resistances, R_2/R_4 .



GB 2 076 547 A

2076547

1/1

FIG.1

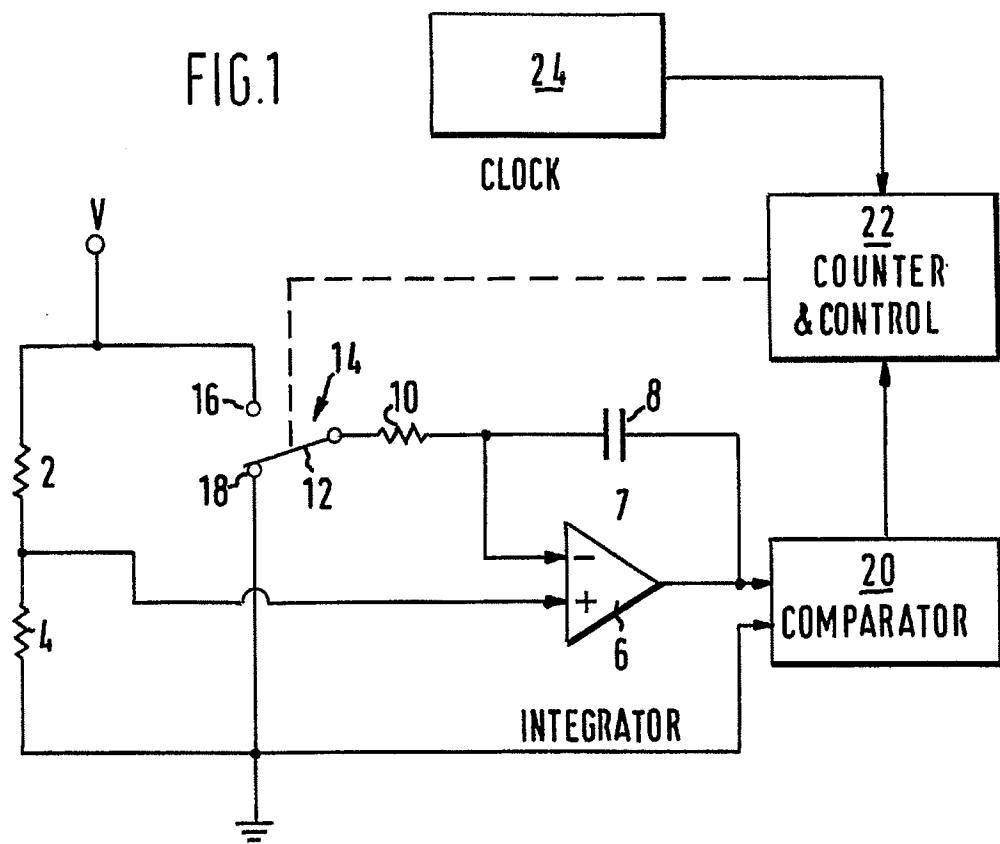
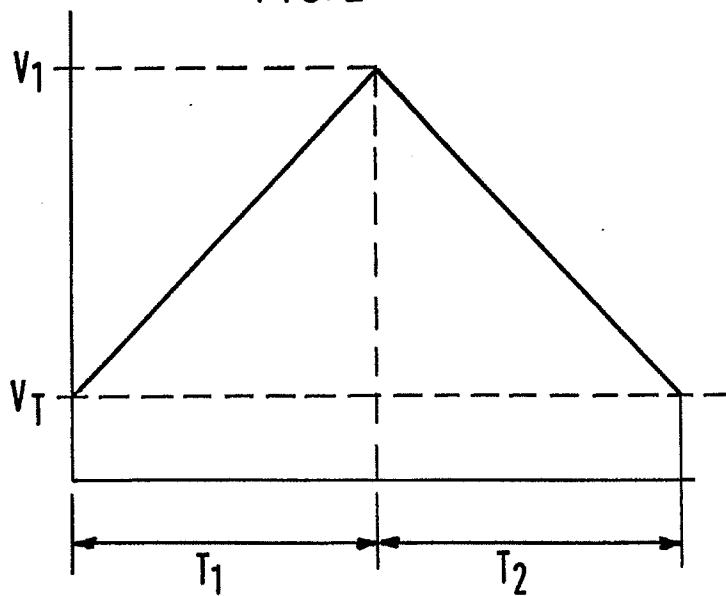


FIG.2



SPECIFICATION
Resistance Measuring Circuit

The present invention is directed to a resistance measuring circuit and more specifically to a circuit that provides a digital representation of the resistance being measured.

5 According to the invention, there is provided a resistance measuring circuit comprising a reference resistance and a resistance to be measured connected in series between a constant voltage source and a circuit ground, an integrator connected to the junction between the resistances, means for producing a first count during a first fixed time period while allowing the integrator to integrate from a fixed threshold level during the first period, and means for producing a second count during a second 10 time period following the first period and having a duration enabling the integrator to return to the fixed 10 threshold level, the circuit being such that the ratio of the first and second counts is proportional to the ratio of the resistance values of the reference and unknown resistances.

An embodiment of the present invention will now be described by way of example only, with reference to the accompanying drawings in which:—

15 15 Figure 1 is a block diagram of an example of a resistance measuring circuit according to the present invention and,

Figure 2 is a waveshape drawing of the operation of the integrator used in the circuit shown in Figure 1.

20 Referring to Figure 1, the resistance measuring circuit includes a known reference resistor 2 and an unknown resistor 4 to be measured connected in series between a supply voltage V and the common or ground connection for the supply voltage V. The resistors 2 and 4, are connected with one end of the reference resistor 2 connected to the supply voltage V and one end of the unknown resistor 4 connected to the common or ground connection. The other ends of the resistors 2, 4 are connected together, and the connection between the series resistors 2 and 4 is connected to the non-inverting 25 input of an operational amplifier 6 arranged in an integrator circuit 7. The output of the amplifier 6, i.e. the output of the integrator 7, is connected to the inverting input of the amplifier 6 through a feedback capacitor 8. The inverting input of the amplifier 6 is also connected through an input resistor 10 to the movable switch arm 12 of a single pole, double throw switch 14. One switch contact 16 of the switch 14 is connected to the supply voltage V while the other switch contact 18 is connected to the common 30 or ground connection.

30 The output of the amplifier 6 is also connected to a first input of a signal comparator 20 which is arranged to compare the output of the integrator 7 with the threshold signal applied to a second input of the comparator 20 from the common or ground connection. The output of the comparator 20 is connected to a control input of a counter and control circuit 22. A clock signal generator, or oscillator, 35 35 40 45 50 55 60 65 70 75 80 85 90 95 100 105 110 115 120 125 130 135 140 145 150 155 160 165 170 175 180 185 190 195 200 205 210 215 220 225 230 235 240 245 250 255 260 265 270 275 280 285 290 295 300 305 310 315 320 325 330 335 340 345 350 355 360 365 370 375 380 385 390 395 400 405 410 415 420 425 430 435 440 445 450 455 460 465 470 475 480 485 490 495 500 505 510 515 520 525 530 535 540 545 550 555 560 565 570 575 580 585 590 595 600 605 610 615 620 625 630 635 640 645 650 655 660 665 670 675 680 685 690 695 700 705 710 715 720 725 730 735 740 745 750 755 760 765 770 775 780 785 790 795 800

The operation of the integrator 7 is illustrated in the waveshape diagram shown in Figure 2. Analyzing the operation of the integrator 7, the initial operation during the first fixed time interval T_1 involves the output of the integrator amplifier 6 ramping to a voltage V_1 , defined by:

$$V_1 = V_T + \frac{1}{RC} \cdot \circ \int_{T_1}^{R_x}{\frac{R_x}{R_x + R_{Ref}} \cdot V dt}$$

55 55 which is equal to:

$$V_1 = V_T + \frac{V}{RC} \cdot \frac{R_x}{R_x + R_{Ref}} \cdot T_1$$

where:

T_1 is the fixed time interval switching of switch 14.

R is the input resistor 10.

V_T is the comparator threshold level.

5 R_x is the unknown resistor R4.

R_{Ref} is the reference resistor R2.

C is the feedback capacitor 8.

The integrator inverting input is then switched to the voltage level of the supply voltage V. The time T_2 required to ramp the output of the amplifier 6 back to the comparator threshold level which is 10 the common or ground level is determined by the equation:

$$V_T = V_1 - \frac{1}{RC} \cdot T_1 + T_2 \int \frac{R_{Ref}}{R_x + R_{Ref}} \cdot V \cdot dt$$

which is equal to

$$V_T = V_1 - \frac{V}{RC} \cdot \frac{R_{Ref}}{R_x + R_{Ref}} \cdot T_2$$

Solving the above equations for

$$\frac{T_2}{T_1} = \frac{R_x}{R_{Ref}}$$

This relationship is independent of the supply voltage V, the resistor 10, the capacitor 8 and the threshold voltage of the comparator and depends only on the ratio of the unknown resistor 4 to the reference resistor 2. The reference resistor 2 is very stable fixed resistor to assure the accuracy of the measurement of the unknown resistor. The ratio of the time intervals,

$$20 \quad \frac{T_2}{T_1} \quad 20$$

is converted to a digital representation by using the fixed and the final count stored in the counter part of the counter and control circuit 22. This conversion does not require a precise clock frequency from the clock circuit 24 since clock frequency must be stable during the time of integration by the integrator amplifier 6. The actual computation of the value of the unknown resistor 4 can be performed 25 by any suitable means, e.g. a microprocessor, etc., such devices being well-known in the art. 25

Claims

1. A resistance measuring circuit comprising a reference resistance and a resistance to be measured connected in series between a constant voltage source and a circuit ground, an integrator connected to the junction between the resistances, means for producing a first count during a first fixed 30 time period while allowing the integrator to integrate from a fixed threshold level during the first period, and means for producing a second count during a second time period following the first period and having a duration enabling the integrator to return to the fixed threshold level, the circuit being such that the ratio of the first and second counts is proportional to the ratio of the resistance values of the reference and unknown resistances.

35 2. The resistance measuring circuit of Claim 1, wherein the integrator includes an operational amplifier; and including a switch for selectively connecting the other input of the amplifier either to said voltage source or said circuit ground; a comparator for comparing the output of the amplifier with a circuit ground level signal; a clock signal generator; and counter and control means for counting the clock signals during said first period while energising the switch to connect the other amplifier input to circuit 40 ground and for energising the switch to connect the other amplifier input to the voltage source during the second period.

3. The resistance measuring circuit of Claim 2, wherein the switch includes a relay controlled by said counter and control means.

4. A resistance measuring circuit substantially as herein described with reference to, and as 45 illustrated in, the accompanying drawings.